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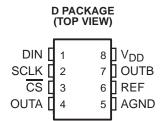
#### features

- **Controlled Baseline** 
  - One Assembly/Test Site, One Fabrication
- **Extended Temperature Performance up to** -55°C to 125°C
- **Enhanced Diminishing Manufacturing** Sources (DMS) Support
- **Enhanced Product Change Notification**
- Qualification Pedigree<sup>†</sup>
- **Dual 12-Bit Voltage Output DAC**
- **Programmable Internal Reference**
- **Programmable Settling Time:** 1 μs in Fast Mode, 3.5 µs in Slow Mode
- † Component qualification in accordance with JEDEC and industry standards to ensure reliable operation over an extended temperature range. This includes, but is not limited to, Highly Accelerated Stress Test (HAST) or biased 85/85, temperature cycle, autoclave or unbiased HAST, electromigration, bond intermetallic life, and mold compound life. Such qualification testing should not be viewed as justifying use of this component beyond specified performance and environmental limits.

- Compatible With TMS320 and SPI™ Serial **Ports**
- Differential Nonlinearity < 0.5 LSB Typ
- **Monotonic Over Temperature**

## applications

- **Digital Servo Control Loops**
- **Digital Offset and Gain Adjustment**
- Industrial Process Control
- **Machine and Motion Control Devices**
- **Mass Storage Devices**



## description

The TLV5638 is a dual 12-bit voltage output DAC with a flexible 3-wire serial interface. The serial interface allows glueless interface to TMS320 and SPI™, QSPI™, and Microwire™ serial ports. It is programmed with a 16-bit serial string containing 4 control and 12 data bits.

The resistor string output voltage is buffered by a x2 gain rail-to-rail output buffer. The buffer features a Class AB output stage to improve stability and reduce settling time. The programmable settling time of the DAC allows the designer to optimize speed vs power dissipation. With its on-chip programmable precision voltage reference, the TLV5638 simplifies overall system design.

Because of its ability to source up to 1 mA, the reference can also be used as a system reference. Implemented with a CMOS process, the device is designed for single supply operation from 2.7 V to 5.5 V. It is available in an 8-pin SOIC package to reduce board space.

#### ORDERING INFORMATION

TA	PACK	\GE‡	ORDERABLE PART NUMBER	TOP-SIDE MARKING
-40°C to 125°C	SOIC - D	Tape and reel	TLV5638QDREP	5638QE
-55°C to 125°C	SOIC - D	Tape and reel	TLV5638MDREP	5638ME

Package drawings, standard packing quantities, thermal data, symbolization, and PCB design guidelines are available at www.ti.com/sc/package.



Please be aware that an important notice concerning availability, standard warranty, and use in critical applications of Texas Instruments semiconductor products and disclaimers thereto appears at the end of this data sheet.

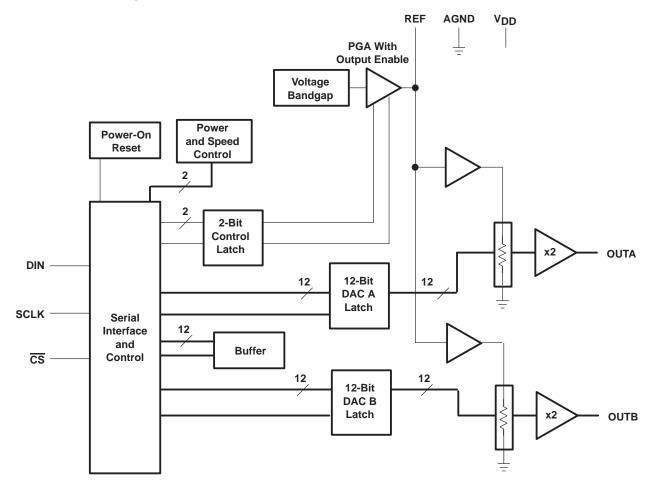
SPI and QSPI are trademarks of Motorola, Inc.

Microwire is a trademark of National Semiconductor Corporation.



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## functional block diagram



## **Terminal Functions**

TERM	INAL	L/O/D	DECORIDATION
NAME	NO.	I/O/P	DESCRIPTION
AGND	5	Р	Ground
CS	3	I	Chip select. Digital input active low, used to enable/disable inputs
DIN	1	I	Digital serial data input
OUT A	4	0	DAC A analog voltage output
OUT B	7	0	DAC B analog voltage output
REF	6	I/O	Analog reference voltage input/output
SCLK	2	I	Digital serial clock input
$V_{DD}$	8	Р	Positive power supply



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# absolute maximum ratings over operating free-air temperature range (unless otherwise noted)†

Supply voltage (V <sub>DD</sub> to AGND)	
Reference input voltage range	
Digital input voltage range	$-0.3 \text{ V to V}_{DD}^{-1} + 0.3 \text{ V}$
Operating free-air temperature range, T <sub>A</sub> : TLV5638Q (see Note 1)	–40°C to 125°C
	–55°C to 125°C
Storage temperature range, T <sub>Stq</sub> (see Note 1)	–65°C to 150°C
Package thermal impedance, θ <sub>JA</sub> : D package	
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds .	260°C

<sup>†</sup> Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

## recommended operating conditions

		MIN	NOM	MAX	UNIT	
Owner have been a M	V <sub>DD</sub> = 5 V	4.5	5	5.5	V	
Supply voltage, V <sub>DD</sub>	V <sub>DD</sub> = 3 V	2.7	3	3.3	V	
Power on reset, POR		0.55		2	V	
	V <sub>DD</sub> = 2.7 V	2			.,	
High-level digital input voltage, VIH	V <sub>DD</sub> = 5.5 V	2.4			V	
	V <sub>DD</sub> = 2.7 V			0.6	.,	
ow-level digital input voltage, V <sub>IL</sub>	V <sub>DD</sub> = 5.5 V			0.8	V	
Reference voltage, V <sub>ref</sub> to REF terminal	V <sub>DD</sub> = 5 V (see Note 1)	AGND	2.048	V <sub>DD</sub> –1.5	V	
Reference voltage, V <sub>ref</sub> to REF terminal	V <sub>DD</sub> = 3 V (see Note 1)	AGND	1.024	V <sub>DD</sub> -1.5	V	
Load resistance, R <sub>L</sub>	·	2			kΩ	
Load capacitance, C <sub>L</sub>				100	pF	
Clock frequency, f <sub>CLK</sub>				20	MHz	
Operating free six temperature T	TLV5638Q (see Note 2)	-40		125	00	
Operating free-air temperature, T <sub>A</sub>	TLV5638M (see Note 2)	-55		125	°C	

- NOTES: 1. Due to the x2 output buffer, a reference input voltage ≥ (V<sub>DD</sub>-0.4 V)/2 causes clipping of the transfer function. The output buffer of the internal reference must be disabled, if an external reference is used.
  - 2. Long term high-temperature storage and/or extended use at maximum recommended operating conditions may result in a reduction of overall device life. See www.ti.com/ep\_quality for additional information on enhanced plastic packaging.



NOTE 1: Long term high-temperature storage and/or extended use at maximum recommended operating conditions may result in a reduction of overall device life. See www.ti.com/ep\_quality for additional information on enhanced plastic packaging.

## TLV5638-EP

## 2.7-V TO 5.5-V LOW-POWER DUAL 12-BIT DIGITAL-TO-ANALOG CONVERTER WITH INTERNAL REFERENCE AND POWER DOWN

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electrical characteristics over recommended operating conditions, V<sub>ref</sub> = 2.048 V, V<sub>ref</sub> = 1.024 V (unless otherwise noted)

## power supply

	PARAMETER	TEST CO	NDITIONS		MIN	TYP	MAX	UNIT	
		No load, All inputs = AGND or V <sub>DD</sub> ,	V <sub>DD</sub> = 5 V,	Fast		4.3	7	A	
			Int. ref.	Slow		2.2	3.6	mA	
	Power supply current		V <sub>DD</sub> = 3 V,	Fast		3.8	6.3	A	
I <sub>DD</sub>			Int. ref.	Slow		1.8	3.0	mA	
			V <sub>DD</sub> = 5 V, Ext. ref.	Fast		3.9	6.3	^	
				Slow		1.8	3.0	mA	
			V <sub>DD</sub> = 3 V,	Fast		3.5	5.7	^	
			Ext. ref.	Slow		1.5	2.6	mA	
	Power-down supply current					0.01	10	μΑ	
DODD	Barrer annah antartar mata	Zero scale, See No	Zero scale, See Note 2			-65		-ID	
PSRR	Power supply rejection ratio	Full scale, See Note 3				-65		dB	

NOTES: 3. Power supply rejection ratio at zero scale is measured by varying VDD and is given by:

 $PSRR = 20 log [(E_{ZS}(V_{DD}max) - E_{ZS}(V_{DD}min))/V_{DD}max]$ 

4. Power supply rejection ratio at full scale is measured by varying V<sub>DD</sub> and is given by:  $PSRR = 20 \log [(E_G(V_{DD}max) - E_G(V_{DD}min))/V_{DD}max]$ 

#### static DAC specifications

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
	Resolution		12			bits
INL	Integral nonlinearity, end point adjusted	See Note 4		±1.7	±6	LSB
DNL	Differential nonlinearity	See Note 5		±0.4	±1	LSB
EZS	Zero-scale error (offset error at zero scale)	See Note 6			±24	mV
E <sub>ZS</sub> TC	Zero-scale-error temperature coefficient	See Note 7		10		ppm/°C
EG	Gain error	See Note 8			±0.6	% full scale V
E <sub>G</sub> T <sub>C</sub>	Gain error temperature coefficient	See Note 9		10		ppm/°C

- NOTES: 5. The relative accuracy or integral nonlinearity (INL) sometimes referred to as linearity error, is the maximum deviation of the output from the line between zero and full scale excluding the effects of zero code and full-scale errors. Tested from code 32 to 4095.
  - 6. The differential nonlinearity (DNL) sometimes referred to as differential error, is the difference between the measured and ideal 1 LSB amplitude change of any two adjacent codes. Monotonic means the output voltage changes in the same direction (or remains constant) as a change in the digital input code.
  - 7. Zero-scale error is the deviation from zero voltage output when the digital input code is zero.

  - Zero-scale-error temperature coefficient is given by: E<sub>ZS</sub> TC = [E<sub>ZS</sub> (T<sub>max</sub>) E<sub>ZS</sub> (T<sub>min</sub>)]/V<sub>ref</sub> × 10<sup>6</sup>/(T<sub>max</sub> T<sub>min</sub>).
     Gain error is the deviation from the ideal output (2V<sub>ref</sub> 1 LSB) with an output load of 10 kΩ excluding the effects of the zero-error.
     Gain temperature coefficient is given by: E<sub>G</sub> TC = [E<sub>G</sub>(T<sub>max</sub>) E<sub>G</sub> (T<sub>min</sub>)]/V<sub>ref</sub> × 10<sup>6</sup>/(T<sub>max</sub> T<sub>min</sub>).



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electrical characteristics over recommended operating conditions,  $V_{ref}$  = 2.048 V,  $V_{ref}$  = 1.024 V (unless otherwise noted) (continued)

## output specifications

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
VO	Output voltage	$R_L = 10 \text{ k}\Omega$	0		V <sub>DD</sub> -0.4	V
	Output load regulation accuracy	$V_{O} = 4.096 \text{ V}, 2.048 \text{ V}, R_{L} = 2 \text{ k}\Omega$			±0.25	% full scale V

## reference pin configured as output (REF)

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
V <sub>ref</sub> (OUTL)	Low reference voltage		1.003	1.024	1.045	V
V <sub>ref</sub> (OUTH)	High reference voltage	V <sub>DD</sub> > 4.75 V	2.027	2.048	2.069	V
I <sub>ref(source)</sub>	Output source current				1	mA
I <sub>ref(sink)</sub>	Output sink current		-1			mA
	Load capacitance				100	pF
PSRR	Power supply rejection ratio			-65		dB

## reference pin configured as input (REF)

	PARAMETER	TEST CONDITIONS	TEST CONDITIONS		TYP	MAX	UNIT
٧ı	Input voltage			0		V <sub>DD-1.5</sub>	V
RĮ	Input resistance				10		ΜΩ
Cl	C <sub>I</sub> Input capacitance			5		pF	
	Reference input bandwidth	PEE 0.2 V + 4.024 V do	Fast		1.3		MHz
	Reference input bandwidth	REF = $0.2 \text{ V}_{pp} + 1.024 \text{ V dc}$	Slow		525		kHz
	Reference feedthrough REF = 1 V <sub>pp</sub> at 1 kHz + 1.024 V dc (see Note 10)			-80	·	dB	

NOTE 11: Reference feedthrough is measured at the DAC output with an input code = 0x000.

## digital inputs

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
lн	High-level digital input current	$V_I = V_{DD}$			1	μΑ
I <sub>IL</sub>	Low-level digital input current	V <sub>I</sub> = 0 V	-1			μΑ
Ci	Input capacitance			8		pF

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# electrical characteristics over recommended operating conditions (unless otherwise noted) (continued)

## analog output dynamic performance

	PARAMETER	TEST	CONDITIONS		MIN	TYP	MAX	UNIT		
	Output as till a self-	$R_L = 10 \text{ k}\Omega$ ,	C <sub>L</sub> = 100 pF,	Fast		1	3	_		
ts(FS)	Output settling time, full scale	See Note 11		Slow		3.5	7	μs		
	Output as till an Cara as da ta sa da	$R_L = 10 \text{ k}\Omega$ ,	C <sub>L</sub> = 100 pF,	Fast		0.5	1.5	_		
ts(CC)	Output settling time, code to code	See Note 12				1	2	μs		
0.0		$R_{I} = 10 \text{ k}\Omega_{I}$ $C_{I} = 100 \text{ p}$		$R_I = 10 \text{ k}\Omega$ , $C_I = 100 \text{ pF}$ , Fast	= 10 kΩ, $C_1$ = 100 pF,	Fast		12		
SR	Slew rate	See Note 13	_	Slow		1.8		V/μs		
	Glitch energy	$\frac{DIN}{CS} = 0 \text{ to } 1,$ $CS = V_{DD}$	FCLK = 100 kF	Hz,		5		nV-s		
SNR	Signal-to-noise ratio				69	74				
S/(N+D)	Signal-to-noise + distortion	f <sub>S</sub> = 480 kSPS,	f <sub>out</sub> = 1 kHz,		58	67				
THD	Total harmonic distortion	$R_L = 10 \text{ k}\Omega$ ,	$C_L = 100 \text{ pF}$			-69	-57	dB		
	Spurious free dynamic range				57	72				

- NOTES: 12. Settling time is the time for the output signal to remain within ±0.5 LSB of the final measured value for a digital input code change of 0x020 to 0xFDF and 0xFDF to 0x020 respectively. Not tested, assured by design.
  - 13. Settling time is the time for the output signal to remain within ± 0.5 LSB of the final measured value for a digital input code change of one count. Not tested, assured by design.
  - 14. Slew rate determines the time it takes for a change of the DAC output from 10% to 90% full-scale voltage.

## digital input timing requirements

		MIN	NOM	MAX	UNIT
t <sub>su(CS-CK)</sub>	Setup time, CS low before first negative SCLK edge	10			ns
tsu(C16-CS)	Setup time, 16 <sup>th</sup> negative SCLK edge (when D0 is sampled) before CS rising edge	10			ns
t <sub>wH</sub>	SCLK pulse width high	25			ns
t <sub>wL</sub>	SCLK pulse width low	25			ns
t <sub>su(D)</sub>	Setup time, data ready before SCLK falling edge	10			ns
th(D)	Hold time, data held valid after SCLK falling edge	5			ns



#### PARAMETER MEASUREMENT INFORMATION

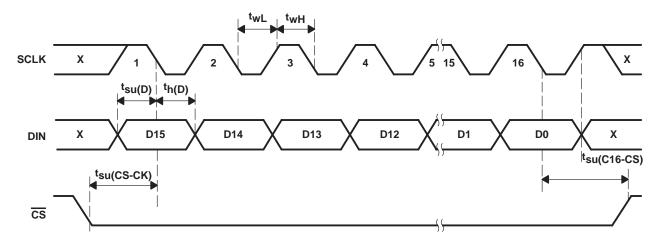
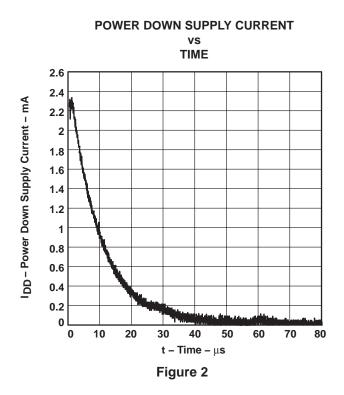
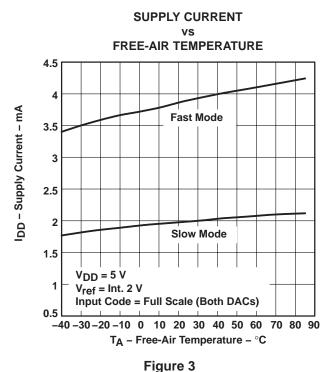


Figure 1. Timing Diagram

#### TYPICAL CHARACTERISTICS





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#### TYPICAL CHARACTERISTICS

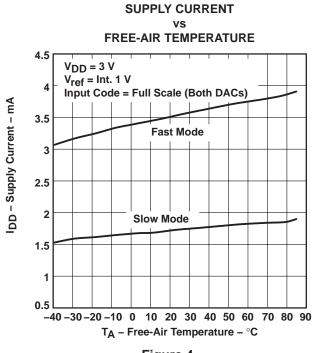
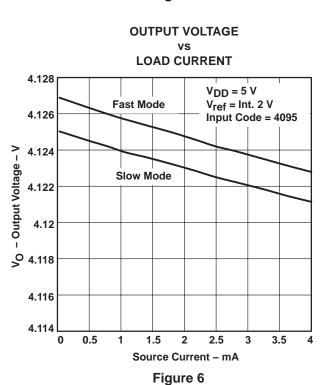


Figure 4



**OUTPUT VOLTAGE LOAD CURRENT** 2.064  $V_{DD} = 3 V$ **Fast Mode**  $V_{ref} = Int. 1 V$ 2.062 Input Code = 4095 - Output Voltage - V 2.06 **Slow Mode** 2.058 2.056 2.054 2.052 2.05 0.5 3 0 1.5 2 2.5 3.5 Source Current - mA

Figure 5

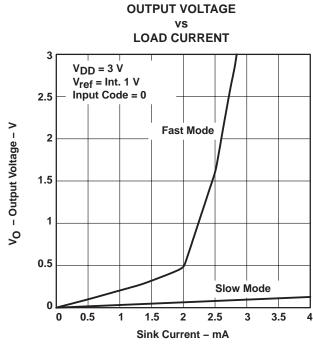
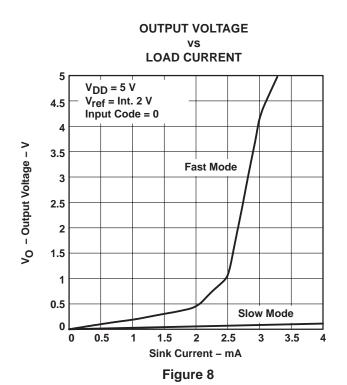
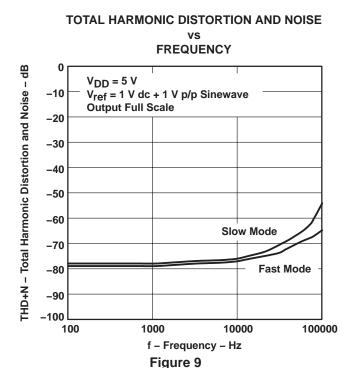


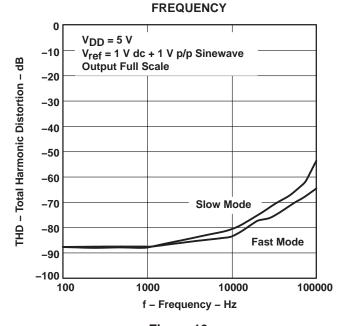
Figure 7

## **TYPICAL CHARACTERISTICS**



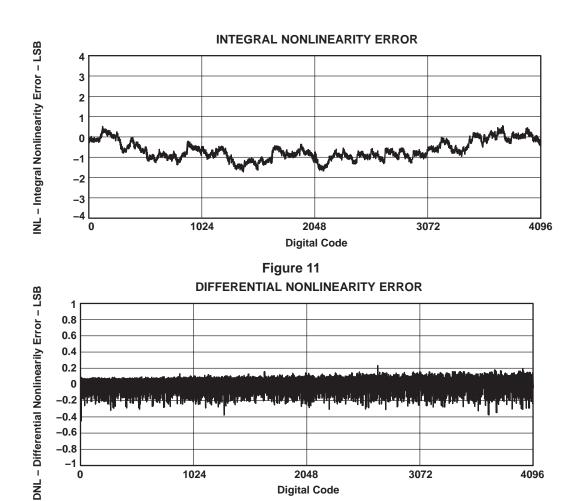


# TOTAL HARMONIC DISTORTION vs





#### TYPICAL CHARACTERISTICS



## **APPLICATION INFORMATION**

Figure 12

## general function

The TLV5638 is a dual 12-bit, single supply DAC, based on a resistor string architecture. It consists of a serial interface, a speed and power-down control logic, a programmable internal reference, a resistor string, and a rail-to-rail output buffer.

The output voltage (full scale determined by reference) is given by:

$$2 \text{ REF } \frac{\text{CODE}}{0 \times 1000} \text{ [V]}$$

Where REF is the reference voltage and CODE is the digital input value in the range 0x000 to 0xFFF. A power on reset initially puts the internal latches to a defined state (all bits zero).



#### **APPLICATION INFORMATION**

#### serial interface

A falling edge of  $\overline{CS}$  starts shifting the data bit-per-bit (starting with the MSB) to the internal register on the falling edges of SCLK. After 16 bits have been transferred or  $\overline{CS}$  rises, the content of the shift register is moved to the target latches (DAC A, DAC B, BUFFER, CONTROL), depending on the control bits within the data word.

Figure 13 shows examples of how to connect the TLV5638 to TMS320, SPI™, and Microwire™.

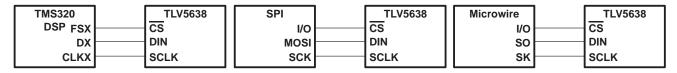


Figure 13. Three-Wire Interface

Notes on SPI™ and Microwire™: Before the controller starts the data transfer, the software has to generate a falling edge on the pin connected to  $\overline{CS}$ . If the word width is 8 bits (SPI™ and Microwire™), two write operations must be performed to program the TLV5638. After the write operation(s), the holding registers or the control register are updated automatically on the 16<sup>th</sup> positive clock edge.

## serial clock frequency and update rate

The maximum serial clock frequency is given by:

$$\rm f_{sclkmax} = \frac{1}{t_{whmin} + t_{wlmin}} = 20 \; MHz$$

The maximum update rate is:

$$f_{updatemax} = \frac{1}{16 (t_{whmin} + t_{wlmin})} = 1.25 \text{ MHz}$$

Note, that the maximum update rate is just a theoretical value for the serial interface, as the settling time of the TLV5638 has to be considered, too.

## **TLV5638-EP**

# 2.7-V TO 5.5-V LOW-POWER DUAL 12-BIT DIGITAL-TO-ANALOG CONVERTER WITH INTERNAL REFERENCE AND POWER DOWN

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#### **APPLICATION INFORMATION**

#### data format

The 16-bit data word for the TLV5638 consists of two parts:

• Program bits (D15..D12)

New data (D11..D0)

	D15	D14	D13	D12	D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
ſ	R1	SPD	PWR	R0						12 Da	ta bits					

SPD: Speed control bit  $1 \rightarrow \text{fast mode}$   $0 \rightarrow \text{slow mode}$ 

PWR: Power control bit  $1 \rightarrow \text{power down}$   $0 \rightarrow \text{normal operation}$ 

The following table lists the possible combination of the register select bits:

#### register select bits

R1	R0	REGISTER							
0	0	Write data to DAC B and BUFFER							
0	1	1 Write data to BUFFER							
1	0	Write data to DAC A and update DAC B with BUFFER content							
1	1	Write data to control register							

The meaning of the 12 data bits depends on the register. If one of the DAC registers or the BUFFER is selected, then the 12 data bits determine the new DAC value:

## data bits: DAC A, DAC B and BUFFER

D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0		
	New DAC Value												

If control is selected, then D1, D0 of the 12 data bits are used to program the reference voltage:

#### data bits: CONTROL

D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
Χ	Х	Х	Х	Х	Х	Х	Х	Х	Х	REF1	REF0

X: don't care

REF1 and REF0 determine the reference source and, if internal reference is selected, the reference voltage.

## reference bits

REF1	REF0	REFERENCE
0	0	External
0	1	1.024 V
1	0	2.048 V
1	1	External

## **CAUTION:**

If external reference voltage is applied to the REF pin, external reference MUST be selected.



#### **APPLICATION INFORMATION**

## **Examples of operation:**

- Set DAC A output, select fast mode, select internal reference at 2.048 V:
  - 1. Set reference voltage to 2.048 V (CONTROL register):

	D15	D14	D13	D12	D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
ſ	1	1	0	1	0	0	0	0	0	0	0	0	0	0	1	0

2. Write new DAC A value and update DAC A output:

D15	D14	D13	D12	D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
1	1	0	0		New DAC A output value										

The DAC A output is updated on the rising clock edge after D0 is sampled.

To output data consecutively using the same DAC configuration, it is not necessary to program the CONTROL register again.

- Set DAC B output, select fast mode, select external reference:
  - 3. Select external reference (CONTROL register):

D15	D14	D13	D12	D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
1	1	0	1	0	0	0	0	0	0	0	0	0	0	0	0

4. Write new DAC B value to BUFFER and update DAC B output:

D15	D14	D13	D12	D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
0	1	0	0				New E	BUFFER	content a	nd DAC	B output	value			

The DAC A output is updated on the rising clock edge after D0 is sampled.

To output data consecutively using the same DAC configuration, it is not necessary to program the CONTROL register again.

- Set DAC A value, set DAC B value, update both simultaneously, select slow mode, select internal reference at 1.024 V:
  - 1. Set reference voltage to 1.024 V (CONTROL register):

D15	D14	D13	D12	D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
1	0	0	1	0	0	0	0	0	0	0	0	0	0	0	1

2. Write data for DAC B to BUFFER:

D15	D14	D13	D12	D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
0	0	0	1						New DAC	B value					

3. Write new DAC A value and update DAC A and B simultaneously:

D15	D14	D13	D12	D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
1	0	0	0					1	New DAC	A value					



#### APPLICATION INFORMATION

## **Examples of operation: (continued)**

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Both outputs are updated on the rising clock edge after D0 from the DAC A data word is sampled.

To output data consecutively using the same DAC configuration, it is not necessary to program the CONTROL register again.

#### Set power-down mode:

D15	D14	D13	D12	D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
Χ	Х	1	Х	Х	Х	Х	Х	Х	Х	Х	Х	Х	Х	Х	Х

X = Don't care

## linearity, offset, and gain error using single ended supplies

When an amplifier is operated from a single supply, the voltage offset can still be either positive or negative. With a positive offset, the output voltage changes on the first code change. With a negative offset, the output voltage may not change with the first code, depending on the magnitude of the offset voltage.

The output amplifier attempts to drive the output to a negative voltage. However, because the most negative supply rail is ground, the output cannot drive below ground and clamps the output at 0 V.

The output voltage then remains at zero until the input code value produces a sufficient positive output voltage to overcome the negative offset voltage, resulting in the transfer function shown in Figure 14.

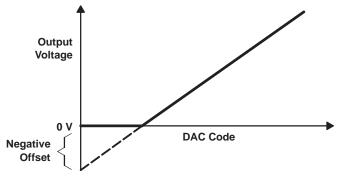


Figure 14. Effect of Negative Offset (Single Supply)

This offset error, not the linearity error, produces this breakpoint. The transfer function would have followed the dotted line if the output buffer could drive below the ground rail.

For a DAC, linearity is measured between zero-input code (all inputs 0) and full-scale code (all inputs 1) after offset and full scale are adjusted out or accounted for in some way. However, single supply operation does not allow for adjustment when the offset is negative due to the breakpoint in the transfer function. So the linearity is measured between full-scale code and the lowest code that produces a positive output voltage.



SGLS130B - JULY 2002 - REVISED DECEMBER 2003

#### APPLICATION INFORMATION

## definitions of specifications and terminology

## integral nonlinearity (INL)

The relative accuracy or integral nonlinearity (INL), sometimes referred to as linearity error, is the maximum deviation of the output from the line between zero and full scale excluding the effects of zero code and full-scale errors.

## differential nonlinearity (DNL)

The differential nonlinearity (DNL), sometimes referred to as differential error, is the difference between the measured and ideal 1 LSB amplitude change of any two adjacent codes. Monotonic means the output voltage changes in the same direction (or remains constant) as a change in the digital input code.

## zero-scale error (EZS)

Zero-scale error is defined as the deviation of the output from 0 V at a digital input value of 0.

#### gain error (E<sub>G</sub>)

Gain error is the error in slope of the DAC transfer function.

### total harmonic distortion (THD)

THD is the ratio of the rms value of the first six harmonic components to the value of the fundamental signal. The value for THD is expressed in decibels.

### signal-to-noise ratio + distortion (S/N+D)

S/N+D is the ratio of the rms value of the output signal to the rms sum of all other spectral components below the Nyquist frequency, including harmonics but excluding dc. The value for S/N+D is expressed in decibels.

## spurious free dynamic range (SFDR)

Spurious free dynamic range is the difference between the rms value of the output signal and the rms value of the largest spurious signal within a specified bandwidth. The value for SFDR is expressed in decibels.







i.com 18-Sep-2008

#### PACKAGING INFORMATION

Orderable Device	Status <sup>(1)</sup>	Package Type	Package Drawing	Pins	Package Qty	e Eco Plan <sup>(2)</sup>	Lead/Ball Finish	MSL Peak Temp <sup>(3)</sup>
TLV5638MDREP	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TLV5638QDREP	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
V62/03628-01XE	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
V62/03628-02XE	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM

<sup>(1)</sup> The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

**NRND:** Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

**OBSOLETE:** TI has discontinued the production of the device.

(2) Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check http://www.ti.com/productcontent for the latest availability information and additional product content details.

TBD: The Pb-Free/Green conversion plan has not been defined.

**Pb-Free** (RoHS): TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.

**Pb-Free (RoHS Exempt):** This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

(3) MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

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#### OTHER QUALIFIED VERSIONS OF TLV5638-EP:

Catalog: TLV5638Military: TLV5638M

NOTE: Qualified Version Definitions:

- Catalog TI's standard catalog product
- Military QML certified for Military and Defense Applications

5-Jul-2008

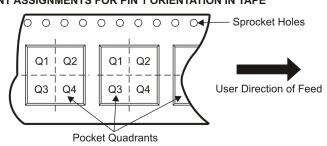
## TAPE AND REEL INFORMATION





	Dimension designed to accommodate the component width
B0	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

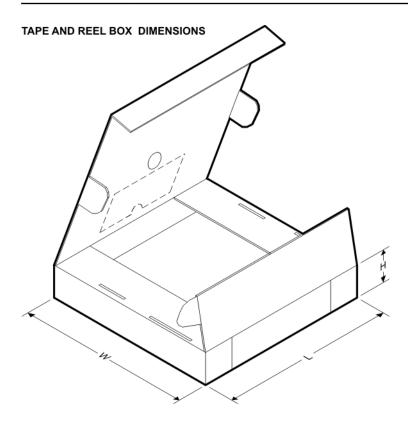
QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



## \*All dimensions are nominal

Device	Package Type	Package Drawing			Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
TLV5638MDREP	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
TLV5638QDREP	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1





\*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
TLV5638MDREP	SOIC	D	8	2500	346.0	346.0	29.0
TLV5638QDREP	SOIC	D	8	2500	346.0	346.0	29.0

# D (R-PDSO-G8)

## PLASTIC SMALL-OUTLINE PACKAGE



NOTES:

- A. All linear dimensions are in inches (millimeters).
- B. This drawing is subject to change without notice.
- Body length does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed .006 (0,15) per end.
- Body width does not include interlead flash. Interlead flash shall not exceed .017 (0,43) per side.
- E. Reference JEDEC MS-012 variation AA.



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